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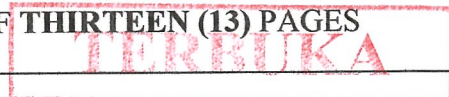


**UNIVERSITI TUN HUSSEIN ONN MALAYSIA**

**FINAL EXAMINATION  
SEMESTER I  
SESSION 2018/2019**

COURSE NAME : SIGNALS AND SYSTEMS  
COURSE CODE : BEB 20203  
PROGRAMME CODE : BEJ  
EXAMINATION DATE : DECEMBER 2018 /JANUARY 2019  
DURATION : 3 HOURS  
INSTRUCTION : SECTION A: ANSWER ALL QUESTIONS  
SECTION B: ANSWER **THREE (3)**  
QUESTIONS ONLY

THIS QUESTION PAPER CONSISTS OF **THIRTEEN (13) PAGES**



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## SECTION A: ANSWER ALL QUESTIONS

**Q1.** (a) According to the given signal  $p(t)$  that shown in **Figure Q1(a)**, sketch a graph for each of the following signals:

(i)  $p(t + 1)$  (2 marks)

(ii)  $-\frac{1}{2}p(0.5t - 1)$  (4 marks)

(b) A system in **Figure Q1(b)** produces an output,  $y(t)$  which is represented as:

$$y(t) = [u(t + 2) - u(t - 2)] + [u(t + 1) - u(t - 1)].$$

By using graphical method, find the input signal  $x(t)$  when  $g(t)$  is given by

$$g(t) = -2[u(t + 1) - u(t - 1)]$$

(4 marks)

**Q2** (a) Consider a periodic signal  $x(t)$  is defined by

$$x(t) = 3 + 4 \sin(2\omega_0 t) + 7 \cos(4\omega_0 t + 30^\circ)$$

(i) Find exponential Fourier Series coefficients of signal  $x(t)$ . (5 marks)

(ii) Sketch the magnitude and phase spectrum of signal  $x(t)$ . (2 marks)

(b) Calculate the average power,  $P_{av}$  supplied to a network if the applied voltage,  $v(t)$  and resulting current,  $i(t)$  are given by

$$v(t) = 50 + 10 \cos\left(30\pi t + \frac{\pi}{2}\right) + 15 \cos\left(45\pi t - \frac{\pi}{3}\right),$$

$$i(t) = 20 + 5 \cos(40\pi t + 20^\circ) - 10 \cos(100\pi t - 50^\circ).$$

(3 marks)

**Q3** (a) A signal  $x(t)$  is given by

$$x(t) = 4 \operatorname{rect}\left(\frac{t}{3}\right).$$

Find the Fourier transform of

(i)  $x(t)$  (2 marks)

(ii)  $y(t) = \frac{d}{dt} (x(t))$  (2 marks)

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(b) Given two signals,

$$x_1(t) = u(t - 1.5) - u(t - 2.5),$$

$$x_2(t) = u(t + 2.5) - u(t + 1.5).$$

(i) Sketch  $y(t) = x_1(t) + x_2(t)$ .

(2 marks)

(ii) Determine  $F[y(t)]$ .

(4 marks)

**Q4.** Given the signal

$$x(t) = 2 \cos(\omega t) u(t).$$

(a) Find the Laplace transform of  $x(t)$  using the definition of Laplace transform.

(4 marks)

(b) Sketch the zero-pole plot and region of convergence (if it exists) of the signal  $x(t)$ .

(2 marks)

(c) Signal  $x(t)$  becomes the input signal for a delayed system as illustrated in **Figure Q4(c)**. Solve  $\mathcal{L}[y(t)]$ .

(4 marks)

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## SECTION B: ANSWER THREE (3) QUESTIONS ONLY

**Q5** (a) Test the stability of the following continuous-time systems:

(i)  $h_1(t) = e^{2t}u(t + 4)$ , (2.5 marks)

(ii)  $h_2(t) = e^{-4t}u(t - 4)$ . (2.5 marks)

(b) Determine and sketch the convolution of the following two signals:

$$x(t) = u(t) - 2u(t - 1) + u(t - 2),$$

$$h(t) = \begin{cases} t, & 0 \leq t \leq 1 \\ 0, & \text{elsewhere.} \end{cases}$$
 (15 marks)

**Q6** (a) Consider the periodic signal  $x(t)$  as shown in **Figure Q6(a)**.

(i) Show that the amplitude-phase Fourier series of  $x(t)$  is

$$x(t) = \frac{1}{3} + \sum_{n=1}^{\infty} \frac{2}{n\pi} \sin \frac{n\pi}{3} \cos \left( \frac{2\pi n}{3 \times 10^{-3}} t \right).$$

(ii) Sketch the amplitude-phase plot for the signal  $x(t)$  for  $n = 0$  to  $n = 5$ .

(14 marks)

(b) The signal  $x(t)$  in **Q6(a)(i)** is passed through a low pass filter (LPF) given in **Figure Q6(b)**.

(i) Determine the frequency response of the LPF. (2 marks)

(ii) If the cutoff frequency,  $f_{cutoff}$  of the LPF is  $f_o$ , where  $f_o$  is the frequency of the first harmonic, determine the output of the LPF for the first harmonic.

(4 marks)

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**Q7** (a) A basic modulator circuit is shown in **Figure Q7(a)**. Modulation is a multiplication between message signal  $m(t)$ , and a carrier signal  $c(t)$ . The process yields a new signal,  $v(t)$ .

(i) Analyze the Fourier transform of the output signal  $v(t)$  by using modulation properties. (4 marks)

(ii) Sketch the spectrum of signal  $v(t)$ . (1 marks)

(b) A Linear Time Invariant (LTI) system has the impulse response,

$$h(t) = \delta(2t + 3).$$

If the input,  $x(t) = e^{-3t}u(t)$ , determine the system output,  $y(t)$ .

(5 marks)

(c) By using the Fourier transform, determine the impulse response,  $h(t)$  for the following system:

$$\frac{dy(t)}{dt} + 2y(t) = x(t) + \frac{dx(t)}{dt}$$

(5 marks)

(d) The voltage across a 10-Ω resistor of an RC circuit in **Figure Q7(d)** is given by  $v_R(t) = 5 e^{-3t}u(t)$  V. Determine the total energy dissipated by this resistor using Parseval's theorem.

(5 marks)

**Q8** (a) The unit impulse response of an LTI system is  $h(t) = e^{-t}(\sin 2t) \cdot u(t)$ . Use the Laplace transform to find its response to

$$x(t) = (\sin(t)) \cdot u(t)$$

(12 marks)

(b) **Figure Q8(b)** shows a transient circuit consists of a DC voltage source, V, a capacitor, C, an inductor, L and a switch. The switch is moved from position 1 to 2 at  $t = 0$ . Determine, using Laplace Transform,

(i) the current flowing through the capacitor before the transition, (1 marks)

(ii) current after the switch is moved from position 1 to 2. (4 marks)

(iii) Sketch the  $i(t)$  if  $V = 5V, R = 10\Omega, C = \frac{1}{10}F$  and  $L = \frac{1}{20}H$  for both before and after the transition.

(3 marks)

-END OF QUESTIONS-

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SEMESTER/SESSION: SEMESTER I/2018/2019  
 COURSE NAME: SIGNALS AND SYSTEMS

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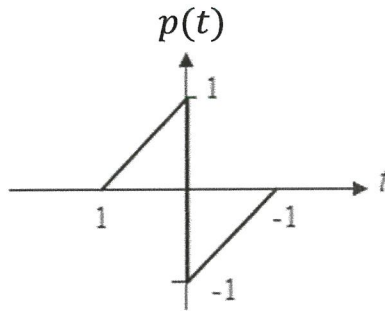


Figure Q1(a)

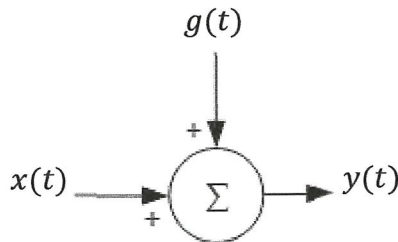


Figure Q1(b)



Figure Q4(c)

FINAL EXAMINATION

SEMESTER/SESSION: SEMESTER I/2018/2019  
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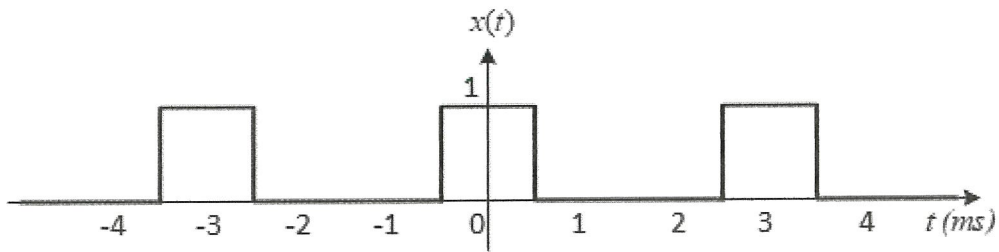


Figure Q6(a)

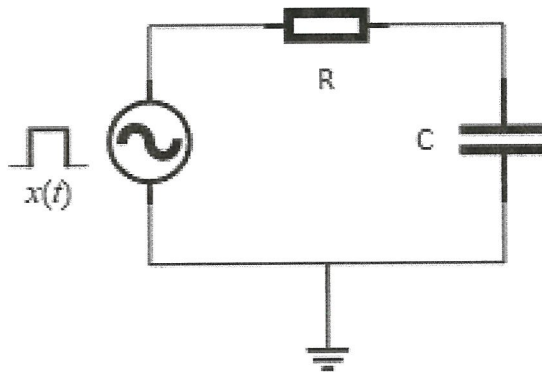


Figure Q6(b)

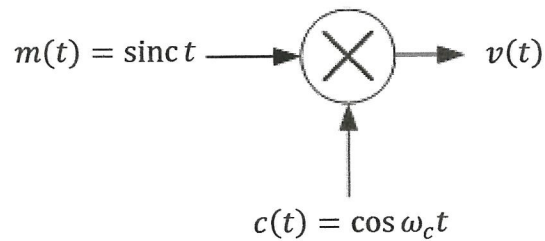


Figure Q7(a)

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SEMESTER/SESSION: SEMESTER I/2018/2019  
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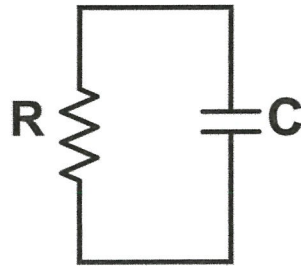


Figure Q7(d)

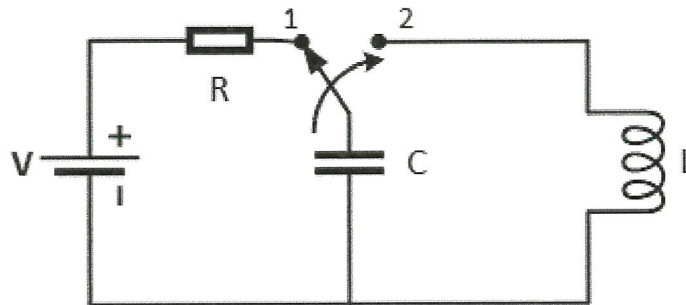


Figure Q8(b)

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SEMESTER/SESSION: SEMESTER I/2018/2019  
 COURSE NAME: SIGNALS AND SYSTEMS

PROGRAMME CODE: BEJ  
 COURSE CODE: BEB 20203

**TABLE 1: INDEFINITE INTEGRALS**

$\int \cos at \, dt = \frac{1}{a} \sin at$	$\int \sin at \, dt = -\frac{1}{a} \cos at$
$\int t \cos at \, dt = \frac{1}{a^2} \cos at + \frac{1}{a} t \sin at$	$\int t \sin at \, dt = \frac{1}{a^2} \sin at - \frac{1}{a} t \cos at$
$\int te^{at} \, dt = \frac{1}{a^2} e^{at} (at - 1)$	$\int \frac{1}{(a^2 + t^2)} \, dt = \frac{1}{a} \tan^{-1} \left( \frac{t}{a} \right)$

**TABLE 2: EULER'S IDENTITY**

$e^{\pm j\pi/2} = \pm j$	$A \angle \pm \theta = Ae^{\pm j\theta}$
$e^{\pm jk\pi} = \cos k\pi$	$e^{\pm j\theta} = \cos \theta \pm j \sin \theta$
$\cos \theta = \frac{1}{2} (e^{j\theta} + e^{-j\theta})$	$\sin \theta = \frac{1}{2j} (e^{j\theta} - e^{-j\theta})$

**TABLE 3: COMPLEX NUMBER**

$s = a + jb =  s  \angle \pm \theta =  s  e^{\pm j\theta}$	$ s  = \sqrt{a^2 + b^2}$	$\theta = \tan^{-1} \left( \frac{b}{a} \right)$
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**TABLE 4: TRIGONOMETRIC IDENTITIES**

$\sin \theta = \cos \left( \theta - \frac{\pi}{2} \right)$	$\cos \theta = \sin \left( \theta + \frac{\pi}{2} \right)$
$\sin(\alpha \pm \beta) = \sin \alpha \cos \beta \pm \cos \alpha \sin \beta$	$\cos(\alpha \pm \beta) = \cos \alpha \cos \beta \mp \sin \alpha \sin \beta$
$\sin^2 \alpha + \cos^2 \beta = 1$	
$\sin 2\alpha = 2 \sin \alpha \cos \alpha$	$\cos 2\alpha = 2 \cos^2 \alpha - 1$
$\cos 2\alpha = 1 - 2 \sin^2 \alpha$	$\cos 2\alpha = \cos^2 \alpha - \sin^2 \alpha$

**TABLE 5: VALUES OF COSINE, SINE AND EXPONENTIAL FUNCTIONS FOR INTEGRAL MULTIPLE OF  $\pi$ .**

Function	Value	Function	Value
$\cos 2n\pi$	1	$e^{j2n\pi}$	1
$\sin 2n\pi$	0	$e^{jn\pi}$	$(-1)^n$
$\cos n\pi$	$(-1)^n$	$e^{\frac{jn\pi}{2}}$	$\begin{cases} (-1)^{\frac{n}{2}}, & n = \text{even} \\ j(-1)^{\frac{n-1}{2}}, & n = \text{odd} \end{cases}$
$\sin n\pi$	0		
$\cos \left( \frac{n\pi}{2} \right)$	$\begin{cases} (-1)^{\frac{n}{2}}, & n = \text{even} \\ 0, & n = \text{odd} \end{cases}$	$\sin \left( \frac{n\pi}{2} \right)$	$\begin{cases} (-1)^{\frac{n-1}{2}}, & n = \text{odd} \\ 0, & n = \text{even} \end{cases}$

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**FINAL EXAMINATION**

SEMESTER/SESSION: SEMESTER I/2018/2019  
 COURSE NAME: SIGNALS AND SYSTEMS

PROGRAMME CODE: BEJ  
 COURSE CODE: BEB 20203

**TABLE 6: FOURIER SERIES**

Exponential	$x(t) = \sum_{n=-\infty}^{\infty} x_n e^{jn\frac{2\pi}{T}t}$ $x_n = \frac{1}{T} \int_{\alpha}^{\alpha+T} x(t) e^{-jn\frac{2\pi}{T}t} dt$
Trigonometric	$x(t) = \frac{a_0}{2} + \sum_{n=1}^{\infty} a_n \cos n\frac{2\pi}{T}t + b_n \sin n\frac{2\pi}{T}t$ $a_n = \frac{2}{T} \int_{\alpha}^{\alpha+T} x(t) \cos n\frac{2\pi}{T}t dt, \quad n = 0, 1, 2, 3 \dots$ $b_n = \frac{2}{T} \int_{\alpha}^{\alpha+T} x(t) \sin n\frac{2\pi}{T}t dt, \quad n = 1, 2, 3 \dots$
Amplitude-phase	$x(t) = X_0 + \sum_{n=1}^{\infty} A_n \cos(n\frac{2\pi}{T}t + \theta_n)$ $A_n = 2 X_n  = \sqrt{a_n^2 + b_n^2}, \quad \theta_n = \angle X_n = -\tan^{-1}\left(\frac{b_n}{a_n}\right)$
Average Power	$P = V_{ac}I_{ac} + \frac{1}{2} \sum_{n=1}^{\infty} V_n I_n \cos(\theta_{V_n} - \theta_{I_n})$

**TABLE 7: DEFINITION OF FOURIER AND LAPLACE TRANSFORM**

<p><b>FOURIER TRANSFORM</b></p> $\mathcal{F}[x(t)] = X(\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t} dt$ $\mathcal{F}[x(t)] = X(f) = \int_{-\infty}^{\infty} x(t)e^{-j2\pi ft} dt$	<p><b>INVERSE FOURIER TRANSFORM</b></p> $x(t) = \mathcal{F}^{-1}[X(\omega)] = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega)e^{j\omega t} d\omega$ $x(t) = \mathcal{F}^{-1}[X(f)] = \int_{-\infty}^{\infty} X(f)e^{j2\pi ft} df$
<p><b>LAPLACE TRANSFORM</b></p> <p><b>Bilateral</b></p> $L[x(t)] = X(s) = \int_{-\infty}^{\infty} x(t)e^{-st} dt$ <p><b>Unilateral</b></p> $L[x(t)] = X(s) = \int_0^{\infty} x(t)e^{-st} dt$ <p><math>s = \sigma + j\omega</math></p>	<p><b>INVERSE LAPLACE TRANSFORM</b></p> $x(t) = L^{-1}[X(s)] = \frac{1}{2\pi j} \int_{c-j\infty}^{c+j\infty} X(s)e^{st} ds$

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FINAL EXAMINATION

SEMESTER/SESSION: SEMESTER I/2018/2019  
 COURSE NAME: SIGNALS AND SYSTEMS

PROGRAMME CODE: BEJ  
 COURSE CODE: BEB 20203

TABLE 8: FOURIER TRANSFORM PAIRS

Time domain, $x(t)$	Frequency domain, $X(\omega)$	Frequency domain, $X(f)$
$\delta(t)$	1	1
1	$2\pi\delta(\omega)$	$\delta(f)$
$u(t)$	$\pi\delta(\omega) + \frac{1}{j\omega}$	$\frac{\delta(f)}{2} + \frac{1}{j2\pi f}$
$u(t + \tau) - u(t - \tau)$	$\frac{2 \sin(\omega\tau)}{\omega} = 2\tau \text{sinc}(\omega\tau)$	$2\tau \text{sinc} 2f\tau$
$\text{rect}(t)$	$\text{sinc}\left(\frac{\omega}{2}\right)$	$\text{sinc}(f)$
$ t $	$-\frac{2}{\omega^2}$	$-\frac{2}{(2\pi f)^2}$
$\text{sgn}(t)$	$\frac{2}{j\omega}$	$\frac{1}{j\pi f}$
$e^{-at}u(t)$	$\frac{1}{a + j\omega}$	$\frac{1}{a + j2\pi f}$
$e^{at}u(-t)$	$\frac{1}{a - j\omega}$	$\frac{1}{a - j2\pi f}$
$e^{-a t }$	$\frac{2a}{a^2 + \omega^2}$	$\frac{2a}{a^2 + 4\pi^2 f^2}$
$e^{j\omega_0 t}$	$2\pi\delta(\omega - \omega_0)$	$\delta(f - f_0)$
$t^n e^{-at}u(t)$	$\frac{n!}{(a + j\omega)^{n+1}}$	$\frac{n!}{(a + j2\pi f)^{n+1}}$
$\sin \omega_0 t$	$\frac{\pi}{j} [\delta(\omega - \omega_0) - \delta(\omega + \omega_0)]$	$\frac{\delta(f - f_0) - \delta(f + f_0)}{2j}$
$\cos \omega_0 t$	$\pi [\delta(\omega + \omega_0) + \delta(\omega - \omega_0)]$	$\frac{\delta(f - f_0) + \delta(f + f_0)}{2}$
$e^{-at} \sin \omega_0 t u(t)$	$\frac{\omega_0}{(a + j\omega)^2 + \omega_0^2}$	$\frac{2\pi f_0}{(a + j2\pi f)^2 + (2\pi f_0)^2}$
$e^{-at} \cos \omega_0 t u(t)$	$\frac{a + j\omega}{(a + j\omega)^2 + \omega_0^2}$	$\frac{a + 2\pi f}{(a + j2\pi f)^2 + (2\pi f_0)^2}$

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**FINAL EXAMINATION**

SEMESTER/SESSION: SEMESTER I/2018/2019  
 COURSE NAME: SIGNALS AND SYSTEMS

PROGRAMME CODE: BEJ  
 COURSE CODE: BEB 20203

**TABLE 9: FOURIER TRANSFORM PROPERTIES**

Property	Time domain, $x(t)$	Frequency domain, $X(\omega)$	Frequency domain, $X(f)$
Linearity	$a_1x_1(t) + a_2x_2(t)$	$a_1X_1(\omega) + a_2X_2(\omega)$	$a_1X_1(f) + a_2X_2(f)$
Time scaling	$x(at)$	$\frac{1}{ a }X\left(\frac{\omega}{a}\right)$	$\frac{1}{ a }X\left(\frac{f}{a}\right)$
Time shifting	$x(t - t_0)u(t - t_0)$	$e^{-j\omega t_0}X(\omega)$	$e^{-j2\pi f t_0}X(f)$
Frequency shifting	$e^{j\omega_0 t}x(t)$	$X(\omega - \omega_0)$	$X(f - f_0)$
Modulation	$\cos(\omega_0 t)x(t)$ $\sin(\omega_0 t)x(t)$	$\frac{1}{2}[X(\omega + \omega_0) + X(\omega - \omega_0)]$ $\frac{1}{2j}[X(\omega - \omega_0) - X(\omega + \omega_0)]$	$\frac{1}{2}[X(f - f_0) + X(f + f_0)]$ $\frac{1}{2j}[X(f - f_0) - X(f + f_0)]$
Time differentiation	$\frac{d}{dt}(x(t))$ $\frac{d^n}{dt^n}(x(t))$	$j\omega X(\omega)$ $(j\omega)^n X(\omega)$	$j2\pi f X(f)$ $(j2\pi f)^n X(f)$
Time integration	$\int_{-\infty}^t x(t)dt$	$\frac{X(\omega)}{j\omega} + \pi X(\omega) \delta(\omega)$	$\frac{X(f)}{j2\pi f} + \frac{1}{2}X(0)\delta(f)$
Frequency differentiation	$t^n x(t)$	$(j)^n \frac{d^n}{d\omega^n} X(\omega)$	$\left(\frac{j}{2\pi}\right)^n \frac{d^n}{df^n} X(f)$
Time Reversal	$x(-t)$	$X(-\omega)$ or $X^*(\omega)$	$X(-f)$
Duality	$X(t)$	$2\pi x(-\omega)$	$X(-f)$
Convolution in $t$	$x_1(t) * x_2(t)$	$X_1(\omega) \cdot X_2(\omega)$	$X(f) \cdot Y(f)$
Multiplication	$x_1(t) \cdot x_2(t)$	$\frac{1}{2\pi} X_1(\omega) * X_2(\omega)$	$X(f) * Y(f)$
Parseval's Theorem	$\int_{-\infty}^{\infty}  x(t) ^2 dt$	$\frac{1}{2\pi} \int_{-\infty}^{\infty}  X(\omega) ^2 d\omega$	$\int_{-\infty}^{\infty}  X(f) ^2 df$

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FINAL EXAMINATION

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TABLE 10: LAPLACE TRANSFORM PAIRS

Time domain, $x(t), t > 0$	s-domain, $X(s)$	ROC	Time domain, $x(t), t > 0$	s-domain, $X(s)$	ROC
$\delta(t)$	1	All $s$	$\cos bt$	$\frac{s}{s^2 + b^2}$	$Re(s) > 0$
$u(t)$	$\frac{1}{s}$	$Re(s) > 0$	$\sin bt$	$\frac{b}{s^2 + b^2}$	$Re(s) > 0$
$t$	$\frac{1}{s^2}$	$Re(s) > 0$	$e^{-at} \cos bt$	$\frac{s+a}{(s+a)^2 + b^2}$	$Re(s) > -a$
$t^n$	$\frac{n!}{s^{n+1}}$	$Re(s) > 0$	$e^{-at} \sin bt$	$\frac{b}{(s+a)^2 + b^2}$	$Re(s) > -a$
$e^{-at}$	$\frac{1}{s+a}$	$Re(s) > -a$	$t \cos bt$	$\frac{s^2 - b^2}{(s^2 + b^2)^2}$	$Re(s) > 0$
$te^{-at}$	$\frac{1}{(s+a)^2}$	$Re(s) > -a$	$t \sin bt$	$\frac{2bs}{(s^2 + b^2)^2}$	$Re(s) > 0$

TABLE 11: LAPLACE TRANSFORM PROPERTIES

Property	Signal	Laplace Transform	ROC
	$x(t)$	$X(s)$	$R$
	$x_1(t), x_2(t)$	$X_1(s), X_2(s)$	$R_1, R_2$
Linearity	$ax_1(t) + bx_2(t)$	$aX_1(s) + bX_2(s)$	At least $R_1 \cap R_2$
Time shifting	$x(t - t_0)$	$e^{-st_0}X(s)$	$R$
Shifting in the s-Domain	$e^{s_0t}x(t)$	$X(s - s_0)$	Shifted version of $R$ (i.e., $s$ is in the ROC if $s - s_0$ is in $R$ )
Time scaling	$x(at)$	$\frac{1}{ a }X\left(\frac{s}{a}\right)$	Scaled ROC (i.e., $s$ is in the ROC if $s/a$ is in $R$ )
Conjugation	$x^*(t)$	$X^*(s^*)$	$R$
Convolution	$x_1(t) * x_2(t)$	$X_1(s) \cdot X_2(s)$	At least $R_1 \cap R_2$
Differentiation in the Time Domain	$\frac{d}{dt}x(t)$	$sX(s)$	At least $R$
		$sX(s) - x(0^+)$ (Unilateral)	$R$ right hand plane
	$\frac{d^n}{dt^n}x(t)$	$s^nX(s) - s^{n-1}x(0^+) - \dots - sx^{n-2}(0^+) - x^{n-1}(0^+)$	
Differentiation in the s-Domain	$-tx(t)$	$\frac{d}{ds}X(s)$	$R$
Integration in the Time Domain	$\int_{-\infty}^t x(\tau)d\tau$	$\frac{1}{s}X(s)$	At least $R \cap \{Re(s) > 0\}$

Initial- and Final- Value Theorems

If  $x(t) = 0$  for  $t < 0$  and  $x(t)$  contains no impulses or higher order singularities at  $t = 0$ , then

$$x(0^+) = \lim_{s \rightarrow \infty} sX(s)$$

If  $x(t) = 0$  for  $t < 0$  and has a finite limit as  $t \rightarrow \infty$ , then

$$\lim_{t \rightarrow \infty} x(t) = \lim_{s \rightarrow 0} sX(s)$$