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**UNIVERSITI TUN HUSSEIN ONN MALAYSIA**

**FINAL EXAMINATION  
SEMESTER I  
SESSION 2016/2017**

COURSE NAME : SIGNALS AND SYSTEMS  
COURSE CODE : BEB 20203  
PROGRAMME : BEJ  
EXAMINATION DATE : DECEMBER 2016/ JANUARY 2017  
DURATION : 3 HOURS  
INSTRUCTION : SECTION A: ANSWER **ALL** QUESTIONS  
SECTION B: ANSWER **THREE (3)**  
QUESTIONS ONLY

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THIS QUESTION PAPER CONSISTS OF **TWELVE (12)** PAGES

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## SECTION A: ANSWER ALL QUESTIONS

Q1 (a) Sketch the signal  $r(t)$  given by

$$r(t) = \frac{3}{5} \left[ 1 - \text{rect} \left( \frac{t}{2} - 3 \right) \right] u(t).$$

(5 marks)

(b) Given a composite signal

$$y(t) = 2 \cos \left( \frac{3\pi}{5} t \right) + \sin \left( \frac{2\pi}{3} t \right),$$

determine the periodicity of  $y(t)$ .

(5 marks)

Q2 (a) Given a signal

$$x(t) = 4 + \cos \left( 3\pi t + \frac{\pi}{3} \right) + 6 \sin \left( 2\pi t + \frac{2\pi}{3} \right)$$

(i) Determine the fundamental angular frequency  $\omega_0$ , of  $x(t)$ .

(3 marks)

(ii) Synthesis the corresponding exponential Fourier Series of  $x(t)$  and its coefficients.

(5 marks)

(b) Using your answer in Q2(a)(iii), determine the average signal power using Parseval's theorem assuming  $R = 1\Omega$ .

(2 marks)

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Q3 (a) Consider the rectangular pulse signal

$$x(t) = \begin{cases} A, & |t| < \frac{\tau}{2} \\ 0, & |t| > \frac{\tau}{2} \end{cases}$$

(i) Derive  $X(\omega)$  using the definition of Fourier Transform.

(3 marks)

(ii) Plot the magnitude spectrum of  $X(\omega)$ .

(4 marks)

(b) Explain the duality property of Fourier Transform with the aid of a simple example.

(3 marks)

Q4 Given the following facts about a real signal  $x(t)$  with Laplace transform of  $X(s)$ .

$X(s)$  has exactly two poles.

$X(s)$  has no zeros in the finite s-plane.

$X(s)$  has a pole at  $s = -1 + j$ .

$X(0) = 8$ .

$e^{2t}x(t)$  is not absolutely integrable.

Determine  $X(s)$  and specify its region of convergence.

(10 marks)

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## SECTION B: ANSWER THREE (3) QUESTIONS ONLY

- Q5** (a)  $x_{even}(t)$  and  $x_{odd}(t)$  are the even and odd components of the signal  $x(t)$  respectively. The signals  $x_{even}(t)u(-t)$  and  $x_{odd}(t)u(-t)$  are given in **Figure Q5(a)(i)** and **Figure Q5(a)(ii)** respectively. Sketch the signal  $x(t)$ .

(5 marks)

- (b) A continuous time signal  $x(t)$  in **Figure Q5(b)** is an input to an LTI system of which the impulse response  $h(t)$  is given by

$$h(t) = u(t).$$

Determine the output  $y(t)$  of the system.

(10 marks)

- (c) **Figure Q5(c)** shows a continuous time systems where  $x(t)$  is the system input and  $y(t)$  is the system output. Deduce the stability of the system  $h(t)$  using suitable justification.

(5 marks)

- Q6** Given a periodic signal  $x(t)$  as shown in **Figure Q6 (a)**.

- (a) Determine the trigonometric Fourier series coefficients ( $a_0$ ,  $a_n$  and  $b_n$ ) of the signal  $x(t)$ .

(10 marks)

- (b) Express the signal  $x(t)$  in trigonometric Fourier series by considering only the first 3 harmonics.

(3 marks)

- (c) The signal  $x(t)$  with the period  $T = 10^{-5}$ s is passed through an LTI system with frequency response given in **Figure Q6 (c)**. Determine the output of the system.

(7 marks)

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- Q7** (a) The Fourier transform of a signal  $x(t)$  is

$$\mathcal{F}[x(t)] = X(\omega) = \text{sinc}\left(\frac{\omega}{2}\right).$$

Using the properties of Fourier transform, determine the Fourier transform of

$$x\left(\frac{t-1}{2}\right) \cos 2t.$$

(4 marks)

- (b) Find the inverse Fourier transform of  $X(\omega) = \frac{1}{(4+j\omega)^2}$  by using convolution property.

(6 marks)

- (c) Determine  $v_0(t)$  of an electrical circuit as shown **Figure Q7(b)** for  $v_i(t) = 3e^{-5t}u(t)$ .

(10 marks)

- Q8** (a) Briefly explain natural response of a system.

(2 marks)

- (b) Consider the circuit in **Figure Q8(b)**. Assume that the current  $i(t)$  has reached a steady state with the switch at position A. At time  $t = 0$ , the switch is moved from position A to position B.

- (i) Find the differential equation relating  $i(t)$  and  $v_2$  for  $t > 0^-$ . Specify the initial condition for the differential equation in terms of  $v_1$ .

(3 marks)

- (c) Given that  $R = 1 \Omega$  and  $L = 1 \text{ H}$  for the circuit in **Figure Q8(b)**, determine and sketch the current  $i(t)$  for each of the following values of  $v_1$  and  $v_2$ ;

- (i)  $v_1 = 0 \text{ V}$ ,  $v_2 = 2 \text{ V}$

(7 marks)

- (ii)  $v_1 = 4 \text{ V}$ ,  $v_2 = 0 \text{ V}$

(4 marks)

- (iii)  $v_1 = 4 \text{ V}$ ,  $v_2 = 2 \text{ V}$

(4 marks)

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- END OF QUESTIONS -

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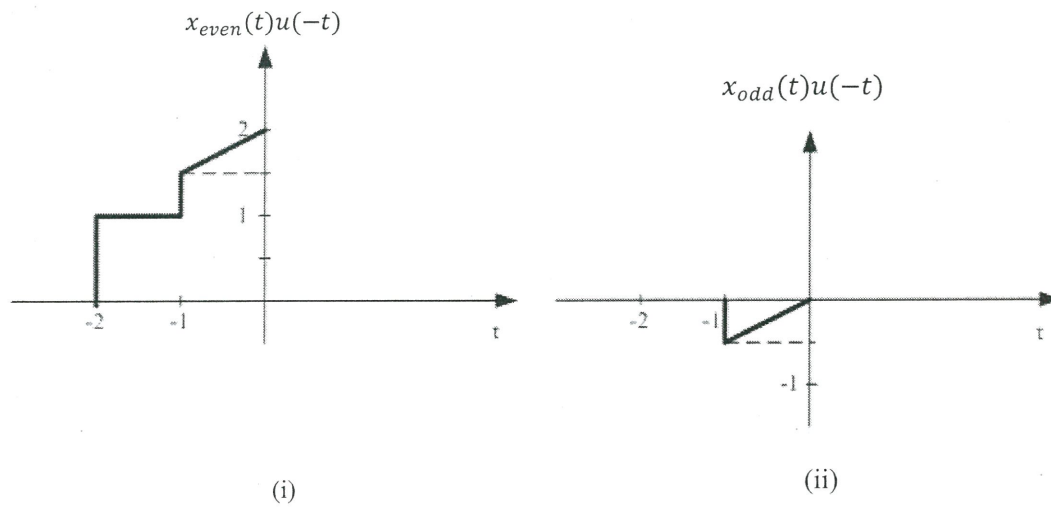


Figure Q5(a)

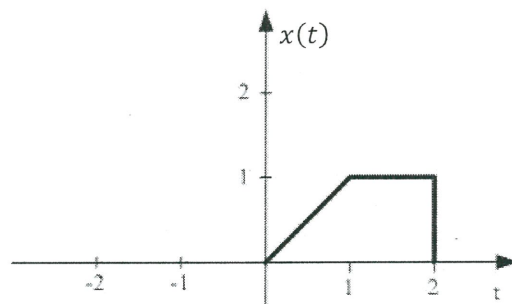


Figure Q5(b)

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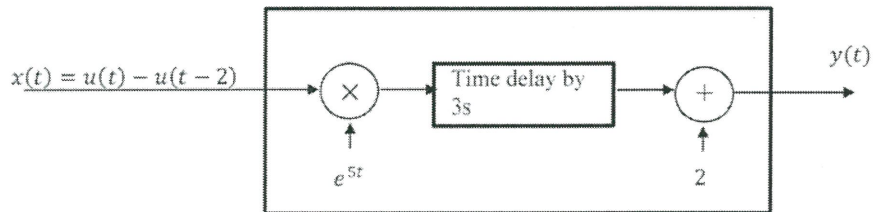


Figure Q5(c)

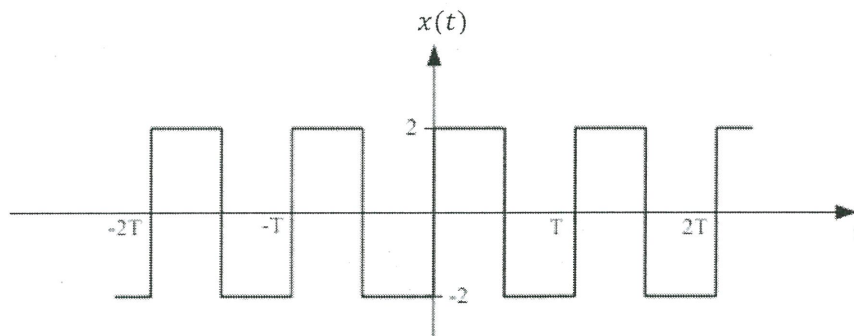


Figure Q6 (a)

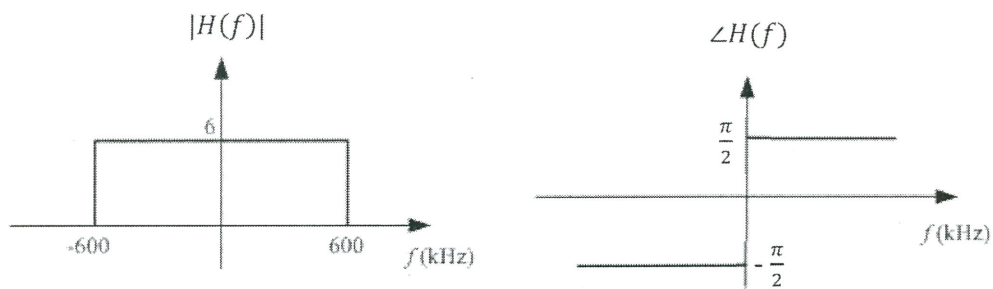


Figure Q6 (c)

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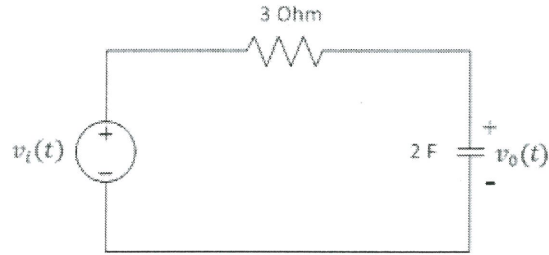


Figure Q7(b)

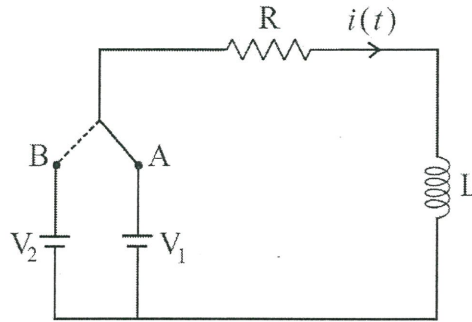


Figure Q8(b)

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**TABLE 1: INDEFINITE INTEGRALS**

$\int \cos at \, dt = \frac{1}{a} \sin at$	$\int \sin at \, dt = -\frac{1}{a} \cos at$
$\int t \cos at \, dt = \frac{1}{a^2} \cos at + \frac{1}{a} t \sin at$	$\int t \sin at \, dt = \frac{1}{a^2} \sin at - \frac{1}{a} t \cos at$
$\int te^{at} \, dt = \frac{1}{a^2} e^{at} (at - 1)$	$\int \frac{1}{(a^2 + t^2)} \, dt = \frac{1}{a} \tan^{-1} \left( \frac{t}{a} \right)$

**TABLE 2: EULER'S IDENTITY**

$e^{\pm j\pi/2} = \pm j$	$A \angle \pm \theta = Ae^{\pm j\theta}$
$e^{\pm jk\pi} = \cos k\pi$	$e^{\pm j\theta} = \cos \theta \pm j \sin \theta$
$\cos \theta = \frac{1}{2} (e^{j\theta} + e^{-j\theta})$	$\sin \theta = \frac{1}{2j} (e^{j\theta} - e^{-j\theta})$

**TABLE 3: COMPLEX NUMBER**

$s = a + jb =  s  \angle \pm \theta =  s  e^{\pm j\theta}$	$ s  = \sqrt{a^2 + b^2}$	$\theta = \tan^{-1} \left( \frac{b}{a} \right)$
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**TABLE 4: TRIGONOMETRIC IDENTITIES**

$\sin \theta = \cos \left( \theta - \frac{\pi}{2} \right)$	$\cos \theta = \sin \left( \theta + \frac{\pi}{2} \right)$
$\sin(\alpha \pm \beta) = \sin \alpha \cos \beta \pm \cos \alpha \sin \beta$	$\cos(\alpha \pm \beta) = \cos \alpha \cos \beta \mp \sin \alpha \sin \beta$
$\sin^2 \alpha + \cos^2 \beta = 1$	
$\sin 2\alpha = 2 \sin \alpha \cos \alpha$	$\cos 2\alpha = 2 \cos^2 \alpha - 1$
$\cos 2\alpha = 1 - 2 \sin^2 \alpha$	$\cos 2\alpha = \cos^2 \alpha - \sin^2 \alpha$

**TABLE 5: VALUES OF COSINE, SINE AND EXPONENTIAL FUNCTIONS FOR INTEGRAL MULTIPLE OF  $\pi$ .**

Function	Value	Function	Value
$\cos 2n\pi$	1	$e^{j2n\pi}$	1
$\sin 2n\pi$	0	$e^{jn\pi}$	$(-1)^n$
$\cos n\pi$	$(-1)^n$	$e^{\frac{jn\pi}{2}}$	$\begin{cases} (-1)^{\frac{n}{2}}, & n = \text{even} \\ j(-1)^{\frac{n-1}{2}}, & n = \text{odd} \end{cases}$
$\sin n\pi$	0		
$\cos \left( \frac{n\pi}{2} \right)$	$\begin{cases} (-1)^{\frac{n}{2}}, & n = \text{even} \\ 0, & n = \text{odd} \end{cases}$	$\sin \left( \frac{n\pi}{2} \right)$	$\begin{cases} (-1)^{\frac{n-1}{2}}, & n = \text{odd} \\ 0, & n = \text{even} \end{cases}$

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**TABLE 6: FOURIER SERIES**

Exponential	$x(t) = \sum_{n=-\infty}^{\infty} x_n e^{jn\frac{2\pi}{T}t}$ $x_n = \frac{1}{T} \int_{\alpha}^{\alpha+T} x(t) e^{-jn\frac{2\pi}{T}t} dt$
Trigonometric	$x(t) = \frac{a_0}{2} + \sum_{n=1}^{\infty} a_n \cos n\frac{2\pi}{T}t + b_n \sin n\frac{2\pi}{T}t$ $a_n = \frac{2}{T} \int_{\alpha}^{\alpha+T} x(t) \cos n\frac{2\pi}{T}t dt, \quad n = 0, 1, 2, 3 \dots$ $b_n = \frac{2}{T} \int_{\alpha}^{\alpha+T} x(t) \sin n\frac{2\pi}{T}t dt, \quad n = 1, 2, 3 \dots$
Amplitude-phase	$x(t) = X_0 + \sum_{n=1}^{\infty} A_n \cos(n\frac{2\pi}{T}t + \theta_n)$ $A_n = 2 X_n  = \sqrt{a_n^2 + b_n^2}, \quad \theta_n = -\tan^{-1}\left(\frac{b_n}{a_n}\right)$

**TABLE 7: DEFINITION OF FOURIER AND LAPLACE TRANSFORM**

<p style="text-align: center;"><b>FOURIER TRANSFORM</b></p> $\mathcal{F}[x(t)] = X(\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t} dt$ $\mathcal{F}[x(t)] = X(f) = \int_{-\infty}^{\infty} x(t)e^{-j2\pi ft} dt$	<p style="text-align: center;"><b>INVERSE FOURIER TRANSFORM</b></p> $x(t) = \mathcal{F}^{-1}[X(\omega)] = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega)e^{j\omega t} d\omega$ $x(t) = \mathcal{F}^{-1}[X(f)] = \int_{-\infty}^{\infty} X(f)e^{j2\pi ft} df$
<p style="text-align: center;"><b>LAPLACE TRANSFORM</b></p> <p style="text-align: center;"><b>Bilateral</b></p> $\mathcal{L}[x(t)] = X(s) = \int_{-\infty}^{\infty} x(t)e^{-st} dt$ <p style="text-align: center;"><b>Unilateral</b></p> $\mathcal{L}[x(t)] = X(s) = \int_0^{\infty} x(t)e^{-st} dt$ <p style="text-align: center;"><math>s = \sigma + j\omega</math></p>	<p style="text-align: center;"><b>INVERSE LAPLACE TRANSFORM</b></p> $x(t) = \mathcal{L}^{-1}[X(s)] = \frac{1}{2\pi j} \int_{c-j\infty}^{c+j\infty} X(s)e^{st} ds$

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**TABLE 8: FOURIER TRANSFORM PAIRS**

Time domain, $x(t)$	Frequency domain, $X(\omega)$	Time domain, $x(t)$	Frequency domain, $X(\omega)$
$\delta(t)$	1	1	$2\pi\delta(\omega)$
$u(t)$	$\pi\delta(\omega) + \frac{1}{j\omega}$	$u(t + \tau) - u(t - \tau)$	$\frac{2 \sin(\omega\tau)}{\omega}$
$ t $	$-\frac{2}{\omega^2}$	$\text{sgn}(t)$	$\frac{2}{j\omega}$
$e^{-at}u(t)$	$\frac{1}{a + j\omega}$	$e^{at}u(-t)$	$\frac{1}{a - j\omega}$
$e^{-a t }$	$\frac{2a}{a^2 + \omega^2}$	$e^{j\omega t}$	$2\pi\delta(\omega - \omega_0)$
$t^n e^{-at}u(t)$	$\frac{n!}{(a + j\omega)^{n+1}}$	$e^{-at} \sin \omega_0 t u(t)$	$\frac{\omega_0}{(a + j\omega)^2 + \omega_0^2}$
$\sin \omega_0 t$	$j\pi[\delta(\omega + \omega_0) - \delta(\omega - \omega_0)]$	$e^{-at} \cos \omega_0 t u(t)$	$\frac{a + j\omega}{(a + j\omega)^2 + \omega_0^2}$
$\cos \omega_0 t$	$\pi[\delta(\omega + \omega_0) + \delta(\omega - \omega_0)]$		

**TABLE 9: FOURIER TRANSFORM PROPERTIES**

Property	Time domain, $x(t)$	Frequency domain, $X(\omega)$
Linearity	$a_1x_1(t) + a_2x_2(t)$	$a_1X_1(\omega) + a_2X_2(\omega)$
Time scaling	$x(at)$	$\frac{1}{ a } X\left(\frac{\omega}{a}\right)$
Time shifting	$x(t - t_0)u(t - t_0)$	$e^{-j\omega t_0} X(\omega)$
Frequency shifting	$e^{j\omega_0 t} x(t)$	$X(\omega - \omega_0)$
Modulation	$\cos(\omega_0 t) x(t)$	$\frac{1}{2} [X(\omega + \omega_0) + X(\omega - \omega_0)]$
	$\sin(\omega_0 t) x(t)$	$\frac{1}{2j} [X(\omega + \omega_0) - X(\omega - \omega_0)]$
Time differentiation	$\frac{d}{dt} x(t)$	$j\omega X(\omega)$
	$\frac{d^n}{dt^n} x(t)$	$(j\omega)^n X(\omega)$
Time integration	$\int_{-\infty}^t x(t) dt$	$\frac{X(\omega)}{j\omega} + \pi X(\omega) \delta(\omega)$
Frequency differentiation	$t^n x(t)$	$(j)^n \frac{d^n}{d\omega^n} X(\omega)$
Time Reversal	$x(-t)$	$X(-\omega)$ or $X^*(\omega)$
Duality	$X(t)$	$2\pi x(-\omega)$
Convolution in $t$	$x_1(t) * x_2(t)$	$X_1(\omega) X_2(\omega)$
Convolution in $\omega$	$x_1(t) x_2(t)$	$\frac{1}{2\pi} X_1(\omega) * X_2(\omega)$

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**TABLE 10: LAPLACE TRANSFORM PAIR**

Time domain, $x(t), t > 0$	s-domain, $X(s)$	ROC	Time domain, $x(t), t > 0$	s-domain, $X(s)$	ROC
$\delta(t)$	1	All $s$	$\cos bt$	$\frac{s}{s^2 + b^2}$	$Re(s) > 0$
$u(t)$	$\frac{1}{s}$	$Re(s) > 0$	$\sin bt$	$\frac{b}{s^2 + b^2}$	$Re(s) > 0$
$t$	$\frac{1}{s^2}$	$Re(s) > 0$	$e^{-at} \cos bt$	$\frac{s+a}{(s+a)^2 + b^2}$	$Re(s) > -a$
$t^n$	$\frac{n!}{s^{n+1}}$	$Re(s) > 0$	$e^{-at} \sin bt$	$\frac{b}{(s+a)^2 + b^2}$	$Re(s) > -a$
$e^{-at}$	$\frac{1}{s+a}$	$Re(s) > -a$	$t \cos bt$	$\frac{s^2 - b^2}{(s^2 + b^2)^2}$	$Re(s) > 0$
$te^{-at}$	$\frac{1}{(s+a)^2}$	$Re(s) > -a$	$t \sin bt$	$\frac{2bs}{(s^2 + b^2)^2}$	$Re(s) > 0$

**TABLE 11: LAPLACE TRANSFORM PROPERTIES**

Property	Signal	Laplace Transform	ROC
	$x(t)$	$X(s)$	$R$
	$x_1(t), x_2(t)$	$X_1(s), X_2(s)$	$R_1, R_2$
Linearity	$ax_1(t) + bx_2(t)$	$aX_1(s) + bX_2(s)$	At least $R_1 \cap R_2$
Time shifting	$x(t - t_0)$	$e^{-st_0} X(s)$	$R$
Shifting in the s-Domain	$e^{s_0 t} x(t)$	$X(s - s_0)$	Shifted version of $R$ (i.e., $s$ is in the ROC if $s - s_0$ is in $R$ )
Time scaling	$x(at)$	$\frac{1}{ a } X\left(\frac{s}{a}\right)$	Scaled ROC (i.e., $s$ is in the ROC if $s/a$ is in $R$ )
Conjugation	$x^*(t)$	$X^*(s^*)$	$R$
Convolution	$x_1(t) * x_2(t)$	$X_1(s) \cdot X_2(s)$	At least $R_1 \cap R_2$
Differentiation in the Time Domain	$\frac{d}{dt} x(t)$	$sX(s)$	At least $R$
		$sX(s) - x(0^+)$ (Unilateral)	$R$ right hand plane
	$\frac{d^n}{dt^n} x(t)$	$s^n X(s) - s^{n-1} x(0^+) - \dots - s x^{n-2}(0^+) - x^{n-1}(0^+)$	
Differentiation in the s-Domain	$-tx(t)$	$\frac{d}{ds} X(s)$	$R$
Integration in the Time Domain	$\int_{-\infty}^t x(\tau) d\tau$	$\frac{1}{s} X(s)$	At least $R \cap \{Re(s) > 0\}$

Initial- and Final- Value Theorems

If  $x(t) = 0$  for  $t < 0$  and  $x(t)$  contains no impulses or higher order singularities at  $t = 0$ , then

$$x(0^+) = \lim_{s \rightarrow \infty} sX(s)$$

If  $x(t) = 0$  for  $t < 0$  and has a finite limit as  $t \rightarrow \infty$ , then

$$\lim_{t \rightarrow \infty} x(t) = \lim_{s \rightarrow 0} sX(s)$$

