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**UNIVERSITI TUN HUSSEIN ONN MALAYSIA**

**FINAL EXAMINATION  
SEMESTER II  
SESSION 2013/2014**

COURSE NAME : DIGITAL SIGNAL PROCESSING  
COURSE CODE : BEB 30503  
PROGRAMME : BEJ  
EXAMINATION DATE : JUN 2014  
DURATION : 3 HOURS  
INSTRUCTION : A) ANSWER **ALL** QUESTIONS  
B) ANSWER **THREE (3)**  
QUESTIONS ONLY

THIS QUESTION PAPER CONSISTS OF **TWELVE (12)** PAGES

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## SECTION A: ANSWER ALL QUESTIONS

- Q1** (a) (i) Analyze signal  $x[n]$  in Figure **Q1(a)** and represent it as a sum of rectangular,  $\text{rect}[n]$ , triangular,  $\text{tri}[n]$ , and impulse,  $\delta[n]$ .

(7 marks)

- (ii) Determine the even and odd symmetry of signal  $f[n] = x[-n+3]$ .

(8 marks)

- (b) Evaluate  $y[n] = x[n-0.5]$  using linear interpolation given,  $x[n] = 6\delta[n+2] + 3\delta[n+1] - 3\delta[n] - 2\delta[n-1] + 4\delta[n-2]$ .

(5 marks)

- Q2** (a) A Linear Time Invariant (LTI) system has TWO (2) impulse response  $h_1[n]$  and  $h_2[n]$  as shown in Figure **Q2(a)**. Evaluate  $y[n]$  using sum by column method if input  $x[n]$  and impulse response  $h_1[n]$  and  $h_2[n]$  given by

$$x[n] = \begin{cases} 2^{|n|} + 1 & -3 < n \leq 1 \\ 0 & \text{elsewhere} \end{cases}$$

$$h_1[n] = 2\delta[n+1] + \delta[n] - \delta[n+1] + 4\delta[n+2] + 3\delta[n+3]$$

$$h_2[n] = 3\delta[n+2] - 2\delta[n+1] + 4\delta[n] + 5\delta[n-1] + \delta[n-2]$$

(10 marks)

- (b) Prove that the cross correlation  $r_{xh}[n] = \frac{0.5^{|n|}}{1-0.5^2}$  if input signal,  $x[n] = 0.5^n u[n]$  and impulse response,  $h[n] = 0.5^n u[n]$ .

(7 marks)

- (c) Analyze the cross correlation obtained from **Q2(b)** for  $0 \leq n \leq 4$ .

(3 marks)

**SECTION B: ANSWER THREE (3) QUESTIONS ONLY**

- Q3** (a) The spectrum of sampled signal of an analog band-limited to some frequency,  $B$  will depend on the sampling frequency,  $S$ . With the aid of diagram, describe the spectrum of the sampled signal for the following condition:
- (i) Oversampling (3 marks)
- (ii) Undersampling (3 marks)
- (b) The analog signal  $x(t) = 2\sin(1000\pi t) + \sin(2000\pi t + \frac{\pi}{2})$  is applied to an analog-to-digital converter (ADC) module for converting the signal to digital format. The ADC has 3 bits quantizer with input voltage range of  $\pm 3V$ .
- (i) Produce the first SIX (6) sampled signals when the sampling frequency is 12.5 kHz. (5 marks)
- (ii) Calculate the quantized signal and encoded digital signal to represent the sampled signal obtained in **Q3(b)(i)** using quantization by rounding. (5 marks)
- (iii) Compute the quantization signal to noise ratio (SNR). (4 marks)
- Q4** (a) Define the N-point discrete Fourier Transform (DFT) of an N-sample signal  $x[n]$  and the inverse DFT (IDFT). (2 marks)
- (b) Given the DFT of  $x[n]$  is  $X_{DFT}[k] = \left\{ \begin{matrix} \downarrow \\ 8, -1-j, -2, -1+j \end{matrix} \right\}$ . Using appropriate properties of the DFT, compute DFT of:
- (i)  $y[n] = x[n-3]$  (2 marks)
- (ii)  $g[n] = x[-n]$  (1 mark)

(iii)  $p[n] = y[n] \otimes g[n]$  (3 marks)

- (c) Calculate the DFT of  $x[n] = \left\{ \begin{matrix} \Downarrow \\ 1, 0, 1, 5 \end{matrix} \right\}$  using Decimation in Time (DIT) Fast Fourier Transform (FFT) approach. (12 marks)

- Q5** (a) Calculate the region of convergence of the following function:

(i)  $\sum_{k=-\infty}^0 \delta[n-k]$  (5 marks)

(ii)  $2^n u[n] - 3^n u[n-1]$  (5 marks)

- (a) Determine the output of the digital system shown in Figure **Q5(b)**, with the input sequence is  $\left\{ \begin{matrix} \Downarrow \\ 3, -1, 3 \end{matrix} \right\}$  and the system is initially at the rest condition. (10 marks)

- Q6** (a) Differentiate equation of  $y[n] - 2y[n-1] + y[n-2] - 2x[n-1] = x[n]$  operates at Nyquist rate of 10 kHz and cutoff frequency of 1 kHz. Design the highpass filter with cutoff frequency of 2 kHz. (10 marks)

- (b) Design Finite Impulse Response (FIR) filter based on structures in Figure **Q6(b)**, with the sampling frequency of 50 Hz and Boxcar windowing order of 5. (10 marks)

- END OF QUESTION -

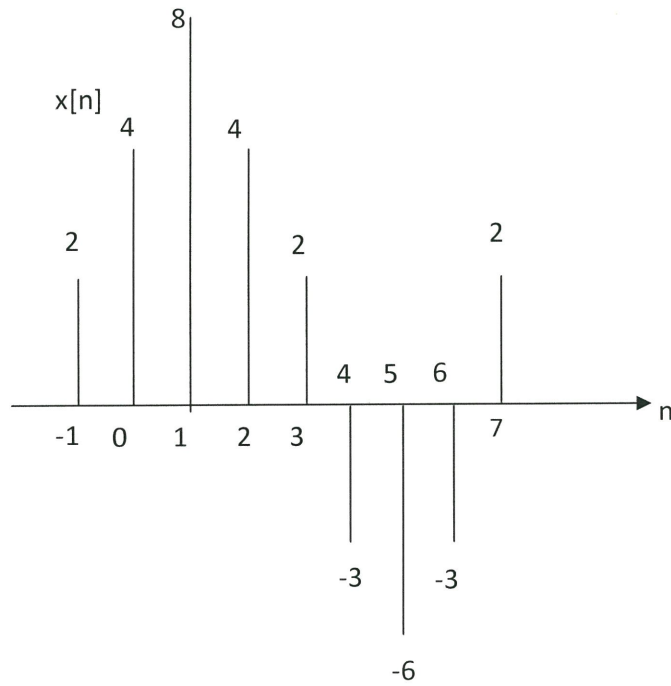
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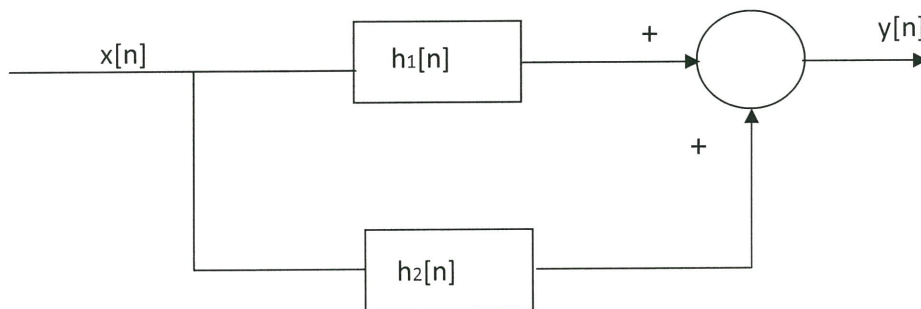
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**FIGURE Q1(a)**



**FIGURE Q2(a)**

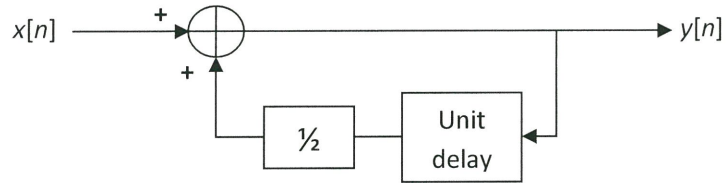
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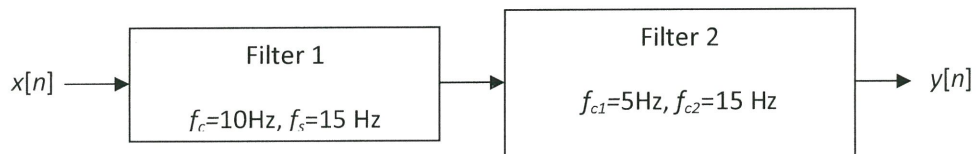
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**FIGURE Q5(b)**



**FIGURE Q6(b)**

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**TABLE 1:** Properties of the  $N$ -Sample DFT

Property	Signal	DFT
Shift	$x[n - n_o]$	$X_{DFT}[k]e^{-j2\pi k n_o/N}$
Shift	$x[n - 0.5N]$	$(-1)^k X_{DFT}[k]$
Modulation	$x[n]e^{j2\pi n k_o/N}$	$X_{DFT}[k - k_o]$
Modulation	$(-1)^n x[n]$	$X_{DFT}[k - 0.5N]$
Folding	$x[-n]$	$X_{DFT}[-k]$
Product	$x[n]y[n]$	$\frac{1}{N} X_{DFT}[k] \otimes Y_{DFT}[k]$
Convolution	$x[n] \otimes y[n]$	$X_{DFT}[k] Y_{DFT}[k]$
Correlation	$x[n] \otimes \otimes y[n]$	$X_{DFT}[k] Y_{DFT}^*[k]$
Central Ordinates	$x[0] = \frac{1}{N} \sum_{k=0}^{N-1} X_{DFT}[k], \quad X_{DFT}[0] = \sum_{n=0}^{N-1} x[n]$	
Central Ordinates	$x\left[\frac{N}{2}\right] = \frac{1}{N} \sum_{k=0}^{N-1} (-1)^k X_{DFT}[k] \quad (N \text{ even}),$ $X_{DFT}\left[\frac{N}{2}\right] = \sum_{n=0}^{N-1} (-1)^n x[n] \quad (N \text{ even})$	

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**TABLE 2:** Properties of z-transform.

Property	Signal	z-transform
Linearity	$a_1x_1(n) + a_2x_2(n)$	$a_1X_1(z) + a_2X_2(z)$
Time reversal	$x(-n)$	$X(z^{-1})$
Time shifting	i) $x(n-k)$ ii) $x(n+k)$	i) $z^{-k}X(z)$ ii) $z^kX(z)$
Convolution	$x_1(n) * x_2(n)$	$X_1(z)X_2(z)$
Correlation	$r_{x_1x_2}(l) = \sum_{n=-\infty}^{\infty} x_1(n)x_2(n-l)$	$R_{x_1x_2}(z) = X_1(z)X_2(z^{-1})$
Scaling	$a^n x(n)$	$X(az)$
Differentiation	$nx(n)$	$z^{-1} \frac{dX(z)}{dz^{-1}}$ or $-z \frac{dX(z)}{dz}$
Time differentiation	$x(n) - x(n-1)$	$X(z)(1 - z^{-1})$
Time integration	$\sum_{k=0}^{\infty} X(k)$	$X(z) = \left( \frac{z}{z-1} \right)$
Initial value theorem	$\lim_{n \rightarrow 0} x(n)$	$\lim_{ z  \rightarrow \infty} X(z)$
Final value theorem	$\lim_{n \rightarrow \infty} x(n)$	$\lim_{ z  \rightarrow 1} \left( \frac{z-1}{z} \right) X(z)$



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**TABLE 3:** Laplace Transform

<b>Signal <math>x(t)</math></b>	<b>Laplace Transform <math>X(s)</math></b>
$\delta(t)$	1
$u(t)$	$\frac{1}{s}$
$r(t) = tu(t)$	$\frac{1}{s^2}$
$t^2u(t)$	$\frac{2}{s^3}$
$t^n u(t)$	$\frac{n!}{s^{n+1}}$
$e^{-\alpha t} u(t)$	$\frac{1}{s + \alpha}$
$te^{-\alpha t} u(t)$	$\frac{1}{(s + \alpha)^2}$
$t^n e^{-\alpha t} u(t)$	$\frac{n!}{(s + \alpha)^{n+1}}$

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**TABLE 4:** Digital to Digital Frequency Transformations

Form	Band Edges	Mapping $z \rightarrow$	Parameters
Lowpass to lowpass	$\Omega_C$	$\frac{z - \alpha}{1 - \alpha z}$	$\alpha = \frac{\sin[0.5(\Omega_D - \Omega_C)]}{\sin[0.5(\Omega_D + \Omega_C)]}$
Lowpass to highpass	$\Omega_C$	$\frac{-(z + \alpha)}{1 + \alpha z}$	$\alpha = \frac{-\cos[0.5(\Omega_D + \Omega_C)]}{\cos[0.5(\Omega_D - \Omega_C)]}$
Lowpass to bandpass	$[\Omega_1, \Omega_2]$	$\frac{-(z^2 + A_1 z + A_2)}{A_2 z^2 + A_1 z + 1}$	$K = \frac{\tan(0.5\Omega_D)}{\tan[0.5(\Omega_2 - \Omega_1)]}$ $\alpha = \frac{-\cos[0.5(\Omega_2 + \Omega_1)]}{\cos[0.5(\Omega_2 - \Omega_1)]}$ $A_1 = \frac{2\alpha K}{K + 1}, A_2 = \frac{K - 1}{K + 1}$
Lowpass to bandstop	$[\Omega_1, \Omega_2]$	$\frac{(z^2 + A_1 z + A_2)}{A_2 z^2 + A_1 z + 1}$	$K = \tan(0.5\Omega_D)\tan[0.5(\Omega_2 - \Omega_1)]$ $\alpha = \frac{-\cos[0.5(\Omega_2 + \Omega_1)]}{\cos[0.5(\Omega_2 - \Omega_1)]}$ $A_1 = \frac{2\alpha}{K + 1}, A_2 = \frac{1 - K}{1 + K}$

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**TABLE 5:** Direct Analog to Digital Transformations for Bilinear Design.

Form	Band Edges	Mapping $s$ →	Parameters
Lowpass to lowpass	$\Omega_C$	$\frac{z-1}{C(z+1)}$	$C = \tan(0.5\Omega_C)$
Lowpass to highpass	$\Omega_C$	$\frac{C(z+1)}{z-1}$	$C = \tan(0.5\Omega_C)$
Lowpass to bandpass	$\Omega_1 < \Omega_0 < \Omega_2$	$\frac{z^2 - 2\beta z + 1}{C(z^2 - 1)}$	$C = \tan[0.5(\Omega_2 - \Omega_1)]$ , $\beta = \cos\Omega_0$ or $\beta = \frac{\cos[0.5(\Omega_2 + \Omega_1)]}{\cos[0.5(\Omega_2 - \Omega_1)]}$
Lowpass to bandstop	$\Omega_1 < \Omega_0 < \Omega_2$	$\frac{C(z^2 - 1)}{z^2 - 2\beta z + 1}$	$C = \tan[0.5(\Omega_2 - \Omega_1)]$ , $\beta = \cos\Omega_0$ or $\beta = \frac{\cos[0.5(\Omega_2 + \Omega_1)]}{\cos[0.5(\Omega_2 - \Omega_1)]}$

**TABLE 6:** Windows for FIR filter design.

Window	Expression $w_N[n]$ , $-0.5(N-1) \leq n \leq 0.5(N-1)$
Boxcar	1
Cosine	$\cos\left(\frac{n\pi}{N-1}\right)$
Riemann	$\text{sinc}^L\left(\frac{2n}{N-1}\right)$ , $L > 0$
Bartlett	$1 - \frac{2 n }{N-1}$
Von Hann (Hanning)	$0.5 + 0.5\cos\left(\frac{2n\pi}{N-1}\right)$
Hamming	$0.54 + 0.46\cos\left(\frac{2n\pi}{N-1}\right)$

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**Finite Summation Formula**

$$\sum_{k=0}^n \alpha = \frac{n(n+1)}{2}$$

$$\sum_{k=0}^n \alpha^2 = \frac{n(n+1)(2n+1)}{6}$$

$$\sum_{k=0}^n \alpha^3 = \frac{n^2(n+1)^2}{4}$$

$$\sum_{k=0}^n \alpha^k = \frac{1-\alpha^{n+1}}{1-\alpha}, \quad \alpha \neq 1$$

$$\sum_{k=0}^n k\alpha^k = \frac{\alpha[1-(n+1)\alpha^n + n\alpha^{n+1}]}{(1-\alpha)^2}$$

$$\sum_{k=0}^n k^2\alpha^k = \frac{\alpha[(1+\alpha)-(n+1)^2\alpha^n + (2n^2+2n-1)\alpha^{n+1} - n^2\alpha^{n+2}]}{(1-\alpha)^3}$$

**Infinite Summation Formula**

$$\sum_{k=0}^{\infty} \alpha^k = \frac{1}{1-\alpha}, \quad |\alpha| < 1$$

$$\sum_{k=1}^{\infty} \alpha^k = \frac{\alpha}{1-\alpha}, \quad |\alpha| < 1$$

$$\sum_{k=1}^{\infty} k\alpha^k = \frac{\alpha}{(1-\alpha)^2}, \quad |\alpha| < 1$$

$$\sum_{k=1}^{\infty} k^2\alpha^k = \frac{\alpha^2 + \alpha}{(1-\alpha)^3}, \quad |\alpha| < 1$$

$$\sum_{k=-\infty}^{\infty} e^{-\alpha|k|} = \frac{1+e^{-\alpha}}{1-e^{-\alpha}}, \quad \alpha > 0$$